

# Filters

\* In this part we will discuss filters design and implementation

## \* Ideal filter Response

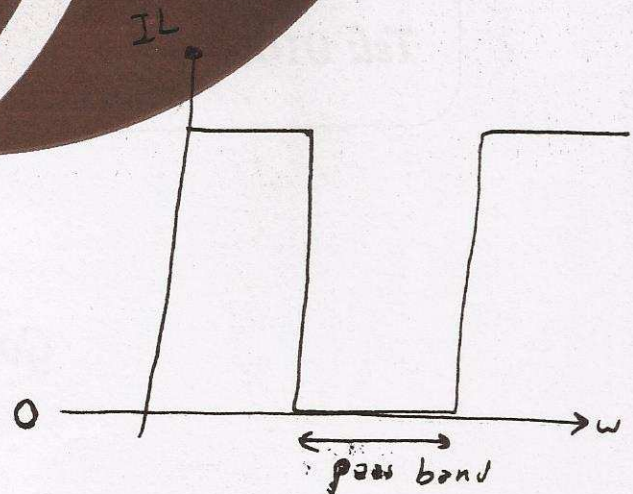
$P_{LR} \equiv$  power loss ratio

$$= \frac{\text{Power available from source}}{\text{Power absorbed by the load}}$$

$$P_{LR} = \frac{1}{1 - |\Gamma(\omega)|^2}$$

$IL \equiv$  The insertion loss

$$IL = 10 \log_{10} P_{LR}$$



Ideal response

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## \* Filter Design techniques:

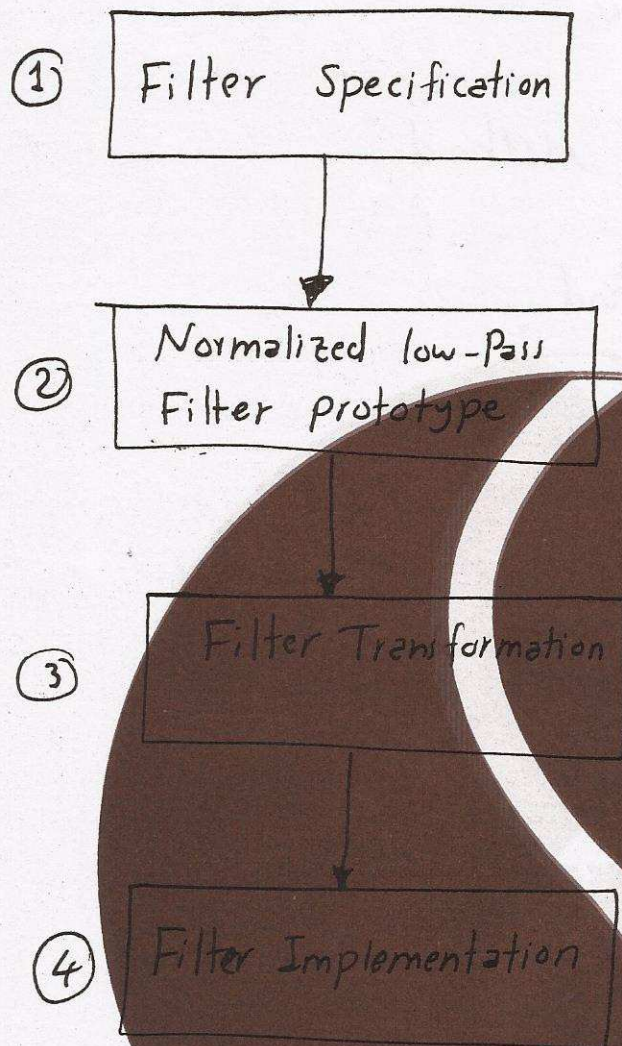
- 1] Periodic structures
- 2] Image Parameter Method
- 3] Insertion loss Method

## \* Filter Implementation techniques:

- 1] Stubs with separating unit Elements
- 2] stepped impedance Low-Pass Filter
- 3] Coupled line filters
- 4] Coupled Resonator Filters.



## \* The flow chart of Realization Procedure :





## I) The Filter Specifications

$$\therefore P_{LR} = \frac{1}{1 - |\Gamma(\omega)|^2} = f(\omega^2)$$

by equating the power loss ratio with different polynomial, we can obtain different filter response

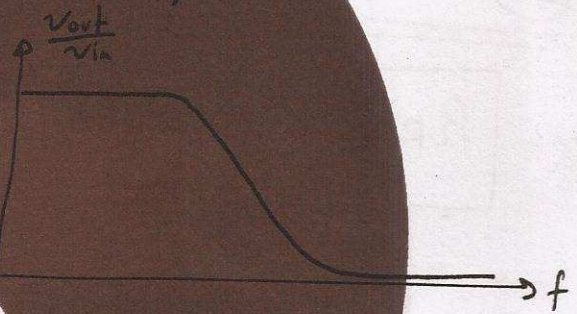
### ① Binomial "Butterworth" "Maximally flat"

$$P_{LR} = 1 + K^2 \left( \frac{\omega}{\omega_c} \right)^{2N}$$

→ maximally flat PB

→ IL in SB "stop band" =  $20N$  dB/decade

→ slow transition



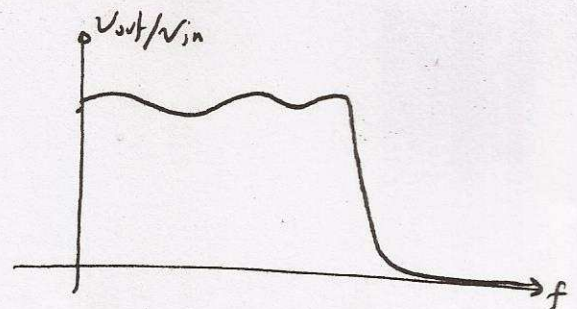
### ② Chebyshev type I

$$P_{LR} = 1 + K^2 T_N^2 \left( \frac{\omega}{\omega_c} \right)$$

→ equal ripple in PB

→ IL in SB  $20N$  dB/decade

→ sharp transition



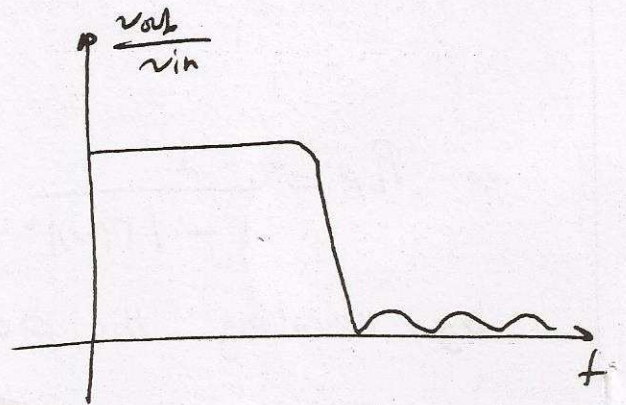
highest IL in SB



### III Inverse chebychev "chebychev type ②"

$$P_{LR} = 1 + \frac{k^2}{T_N^2\left(\frac{\omega_c}{\omega}\right)}$$

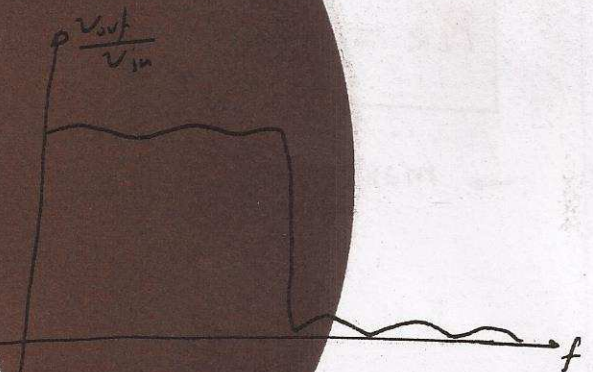
- flat in PB
- Equal ripple in SB
- sharp Transition



### IV Elliptic

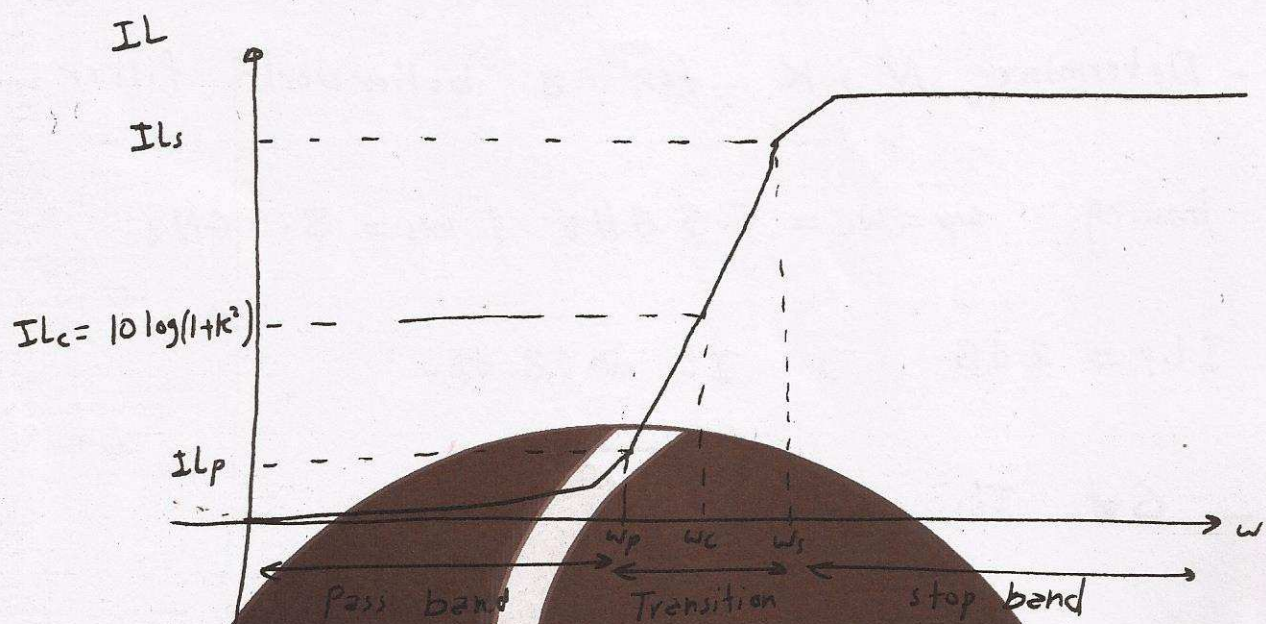
$$P_{LR} = 1 + k^2 R_N^2\left(\frac{\omega}{\omega_c}\right)$$

- equal ripple in PB
- equal ripple in SB
- sharpest transition





## \* low pass filter prototype response



Given  $IL_p$ ,  $IL_s \Rightarrow$  we need to calculate

$k$ ,  $N$  [  $N \equiv$  the filter order ]

There are two Methods to obtain  $N, k$

① Analytically

② Graphically



### Example

- Determine  $N, K$  for a Butterworth filter

having  $\omega_p = \omega_c = 5.5 \text{ GHz}$ ,  $\omega_s = 8.9 \text{ GHz}$

$$IL_p = 2 \text{ dB}, \quad IL_s \geq 18 \text{ dB}$$

- Get  $IL_s$  for your filter

⊕ We will use the analytical solution

$$|P_{LR}| = 1 + K^2 \left( \frac{\omega}{\omega_c} \right)^{2N} \rightarrow \text{Binomial}$$

$$\star IL_p = 10 \log |P_{LR}|_{\omega=\omega_p=\omega_c} = 2 \text{ dB}$$

$$IL_p = 10 \log (1 + K^2) = 2 \text{ dB}$$

$$1 + K^2 = 10^{0.2}$$

$$K = \sqrt{10^{0.2} - 1} = 0.764783$$

$$\star IL_s = 10 \log |P_{LR}|_{\omega=\omega_s} \geq 18 \text{ dB}$$



$$10 \log \left[ 1 + K^2 \left( \frac{\omega_s}{\omega_c} \right)^{2N} \right] \geq 18 \text{ dB}$$

$$\left[ 1 + K^2 \left( \frac{\omega_s}{\omega_c} \right)^{2N} \right] \geq 10^{1.8}$$

$$K^2 \left( \frac{\omega_s}{\omega_c} \right)^{2N} \geq 10^{1.8} - 1$$

$$\left( \frac{\omega_s}{\omega_c} \right)^{2N} \geq \left[ \frac{10^{1.8} - 1}{K^2} \right]$$

$$2N \ln \left( \frac{\omega_s}{\omega_c} \right) \geq \ln \left[ \frac{10^{1.8} - 1}{K^2} \right]$$

$$N \geq \frac{\ln \left[ (10^{1.8} - 1) / K^2 \right]}{2 \ln(\omega_s / \omega_c)}$$

$$N \geq 4.84622$$

$$\boxed{N = 5} \leftarrow \text{next integer}$$

لاحظ اذا ذكر في المسألة  $\Delta \text{dB} = 18$

هذه المقصود  $\Delta \text{dB} \geq 18$  ان نأخذ تحقق القيمة

المطلوب أو أكثر منها

لأنه 11 (filter order) يجب أن يكون integer



$$\therefore IL_s = 10 \log \left[ 1 + k^2 \left( \frac{\omega_1}{\omega_c} \right)^{2N} \right]$$

$$= 10 \log \left[ 1 + (0.766783)^2 \times \left( \frac{8.9}{5.5} \right)^{10} \right]$$

$$IL_s = 18.633399 \text{ dB}$$

ہم قنا بتحقق رلا اکبر سے المطرب





### Example

- Determine  $N, K$  for a butter worth filter  
having  $\omega_c = 5.5 \text{ GHz}$  ,  $\omega_s = 8.9 \text{ GHz}$

$$IL_s = 18 \text{ dB}$$

لا ط في المبدأ أن الساتبة تم إعطاء معلومتين  $IL_p$  و  $IL_s$

حتى نحصل على  $K$  و  $N$

أما في هذا المثال فإن لدينا معلومة واحدة فقط  $IL_s = 18 \text{ dB}$

هذه في هذه الحالة نفترض دائما أن هناك معلومة أخرى  $IL_p = 3 \text{ dB}$

$$\Rightarrow IL_p = 10 \log \left[ 1 + K^2 \left( \frac{\omega_c}{\omega_p} \right)^{2N} \right] = 3$$

$$= 10 \log [1 + K^2] = 3$$

$$\Rightarrow \boxed{K=1}$$

$$IL_s = 10 \log \left[ 1 + \left( \frac{\omega_s}{\omega_c} \right)^{2N} \right] = 18$$

$$1 + \left( \frac{\omega_s}{\omega_c} \right)^{2N} = 10^{1.8}$$

$$2N \ln \left( \frac{\omega_s}{\omega_c} \right) = \ln [10^{1.8} - 1]$$



$$N = \frac{\ln[10^{1.8} - 1]}{2 \ln(\omega_s/\omega_c)} = 4.289$$

$$N = 5$$

Analytical solution

کل ما سبق کا ار

II Graphical solution

Filter order (N)

لا بچار (N)

curves

مجموعہ میں

یہ استفادہ

curves مختلف

Filter

کل نوع

\* Binomial "Butterworth" → has its own curves

\* Chebyshev type 1 → has its own curves

دیکھا

← لاحظہ کل نوع Filter یہ رسمہ اکثر سے مرہ → سبقہ ار K

فہملا ار curves کے صفحہ الی

① Butterworth

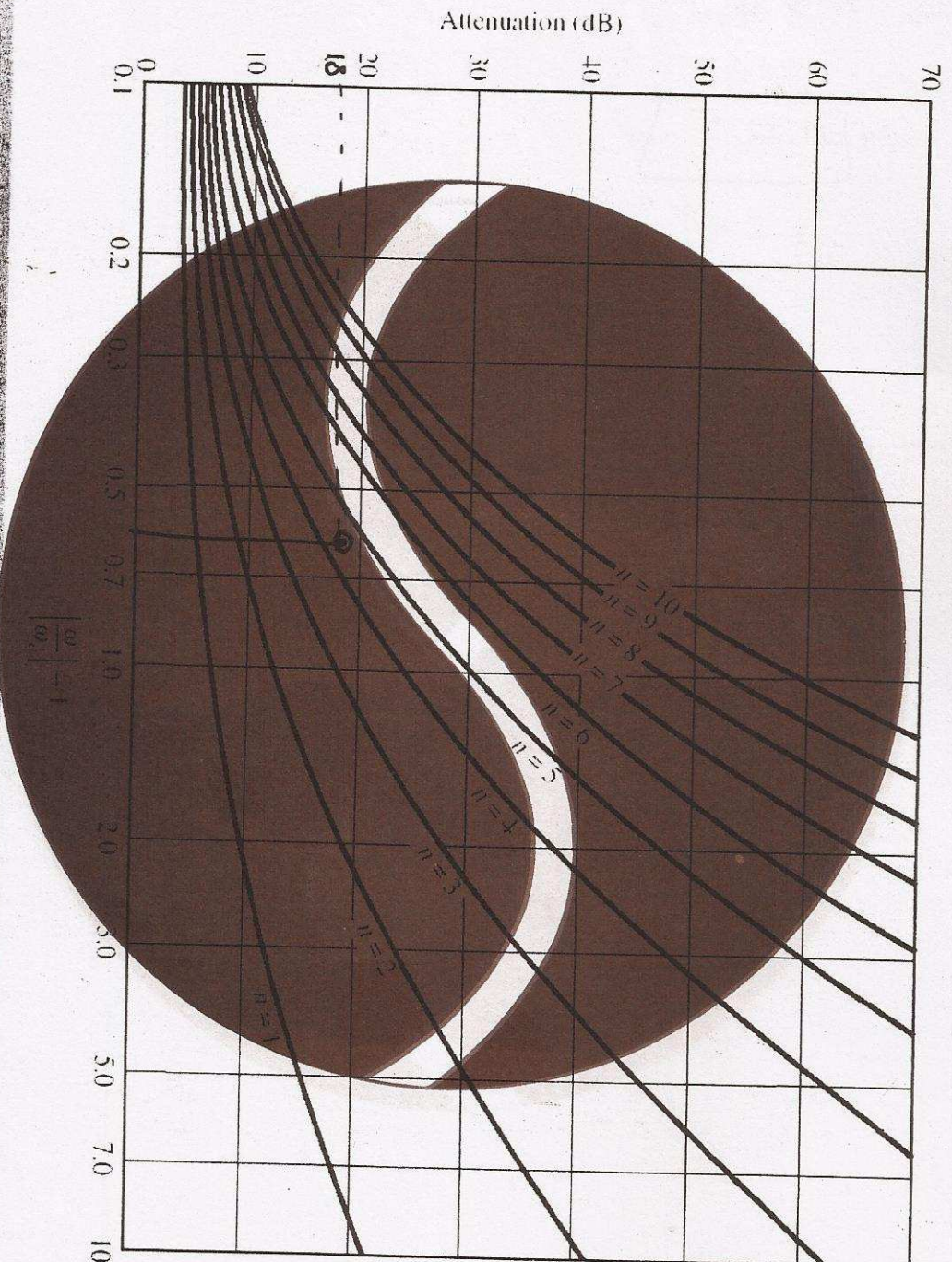
② K = 1 (3 dB at  $\omega = \omega_c$ )

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# Low-Pass Filter Prototype Design Curves for Maximally Flat LPF



Given the attenuation (insertion loss) at a frequency in the stopband frequency, the order of the filter  $N$  can be determined.

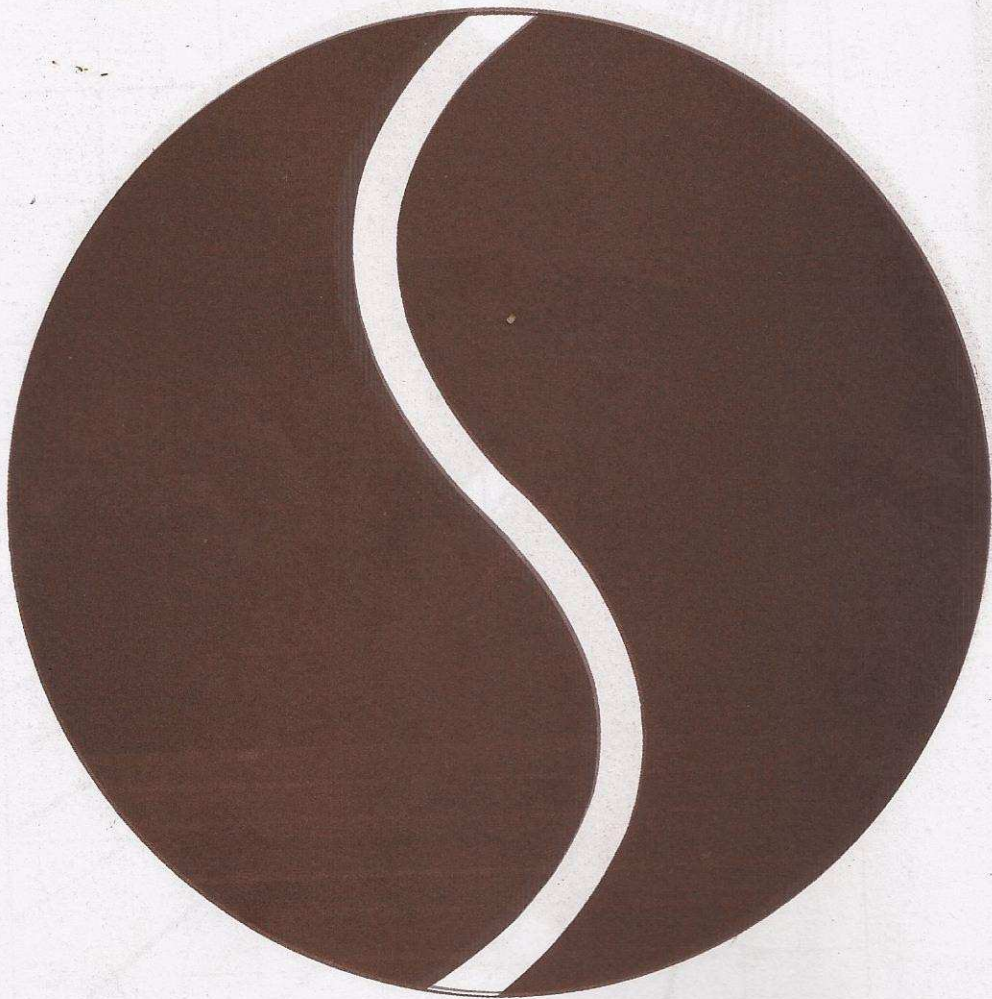


As indicated in the previous graph

$$\therefore \left| \frac{\omega_1}{\omega_c} \right| - 1 = \frac{8.9}{5.5} - 1 = \underline{\underline{0.618}}$$

$$\Rightarrow n > 4$$

$$\therefore \boxed{n = 5}$$





## 2 Normalized low-Pass filter Prototype

- \* Given the filter order  $N$
- \* and the filter type [Butterworth, chebyshev....]
- \* the value of  $k$

→ We need to get the value of the capacitors and inductors used in the prototype

\* There are two types of solutions

1 Analytical

2 Using tables

Hint → to get the prototype capacitors and inductors value we always assume the following assumptions

1  $R_s = 1$

2  $\omega_c = 1$



### Exempl

Get the prototype capacitance and inductor values for a Butterworth filter of 2<sup>nd</sup> order

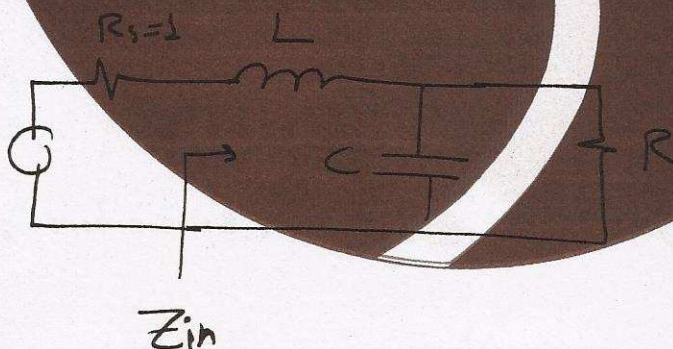
⊖ Analytical solution

$$K=1$$

for butterworth only

2<sup>nd</sup> order filter  $\Rightarrow N=2$

The prototype can be plotted as follow



$$Z_{in} = j\omega L + \left( \frac{1}{j\omega C} \parallel R \right)$$